An Area Efficient FFT Implementation for OFDM

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Abstract— Multiplier less Inverse Fast Fourier Transform is a efficient technique for designing an orthogonal frequency division multiplexing based digital transmitter. OFDM plays a vital role in wireless systems like IEEE 802.11x and IEEE 802.16x. In this paper, an efficient way of FFT implementation has to be done. This works mainly concentrates on reduction of hardware complexity and power consumption due to the complex multiplication involved in twiddle factors at the input stage of IFFT processing unit. This trivial multiplication can be replaced by concepts of 'pass-logic' which eliminates the multiplication factor in butterfly module. This technique optimises the FFT structure with less power and less area. The comparison of various N point FFT with multipliers and pass-logic has been tabulated. The performance improvement of FFT has to be analysed. The proposed architecture with pass logic is an encouraging phenomenon in OFDM based wireless systems.

Key words: FFT, Inverse Fast Fourier Transform, OFDM, passlogic, twiddle multiplier

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INTRODUCTION

Fast Fourier Transform plays a vital role in a modern digital signal processing and telecommunications, especially in orthogonal frequency division multiplexing systems, such as wired and wireless systems. Orthogonal frequency division multiplexing (OFDM) will suitable for video transmission and mobile Internet applications. OFDM techniques have received great attention in high-speed data communication systems and have been selected for wireless local area network IEEE 802.11a and Hiperlan-2, digital audio broadcasting, digital video broadcasting, very high-speed digital subscriber line, and beyond 3G research. In general, FFT and IFFT are used as a digital transmitter and receiver in the OFDM systems.

The computational complexity in FFT processor unit increasing as the number of sample point increases. Therefore, the practical implementation of FFT limits the power consumption, silicon area and bit processing rate. The efficiency of FFT unit is improved by redesigning the arithmetic units, modifying the structure and adopting the modulation aware multipliers, choosing the pipeline strategy and redesigning the memory controller. A novel pass-logic technique is introduced to reduce area, power and to improve the performance. The proposed method simplifies the hardware complexity at the input stage of Npoint inverse FFT structure. This paper describes about the implementation of IFFT technique with pass logic and multipliers and comparison of various parameters have been analyzed. The results produced have been compared with normal FFT multiplier techniques. This comparison shows pass logic technique is the better one in terms of power, area and delay. However, the basic FFT structure cannot provide the better performance due to computationally intensive tasks.

A. RADIX-2 DIF FFT ALGORITHM

The most basic FFT algorithm is the Radix-2 Decimation in frequency algorithm. This algorithm decomposes even and odd-indexed frequency samples shown mathematically in equation set as

$$\begin{aligned} x(2k) &= \sum_{n=0}^{N-1} [x(n)W_N^{2kn}] \\ &= \sum_{n=0}^{\frac{N}{2}-1} [x(n)W_N^{2kn}]_+ \\ &\sum_{n=0}^{\frac{N}{2}-1} [x(n+\frac{N}{2})W_N^{2k(n+\frac{N}{2})}] \\ &= \sum_{n=0}^{\frac{N}{2}-1} [x(n)W_N^{2kn}]_+ \\ &\sum_{n=0}^{\frac{N}{2}-1} [x(n+\frac{N}{2})W_N^{2kn}] \\ &= \sum_{n=0}^{\frac{N}{2}-1} [(x(n)+x(n+\frac{N}{2}))W_N^{kn}] \\ &= DFT_{\frac{N}{2}} [x(n)+x(n+\frac{N}{2})] \\ &x(2k+1) = \sum_{n=0}^{N-1} [x(n)W_N^{(2k+1)n}] \\ &= \sum_{n=0}^{\frac{N}{2}-1} [(x(n)+W_N^{\frac{N}{2}}x(n+\frac{N}{2}))W_N^{(k+1)n}] \end{aligned}$$

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$$= \sum_{n=0}^{\frac{N}{2}-1} \left[(x(n) - x(n + \frac{N}{2})) W_N^n W_{\frac{N}{2}}^{kn} \right]$$
$$= DFT_{\frac{N}{2}} \left[(x(n) - x(n + \frac{N}{2})) W_N^n \right]$$

The decimation in frequency FFT algorithm decomposes the DFT by recursively splitting the sequence elements X (k) in the frequency domain into smaller and smaller sub sequences. The algorithm carries complex addition and complex multiplication operation. These computations are done in radix 2 butterfly structure as shown in figure. This is similar to DIT algorithm except the inputs are given in normal order and outputs are taken in bit reversal order and twiddle multipliers are multiplied after the addition operation.

B. Design of DIF FFT architecture with pass logic

In OFDM system N-point radix-2 Inverse FFT is implemented on transmitter side which has n stages. This architecture has maximum number of twiddle multipliers at the input stage (nth stage). The inputs at the nth stage are converted into 'w' bits which follows any one of the value $\{0, \dots, 0\}$ TF,-TF}, where TF represents twiddle factor. The proposed pass-logic passes inputs as $\{0, 1, -1\}$ and TF, which can replaces the twiddle multiplication at the nth stage of each butterfly structure. In R2-DIF-FFT architecture represented in Fig.1used in the OFDM based transceiver shows the number of frequency divisions with respect to complexity in N point FFT. The complexity of N point FFT can be presented in the Table I describes the number of stages, (N/2 log₂ N) total multipliers and there are 2ⁿ-1 number of multipliers used in nth stage of FFT unit. The hardware complexity would certainly reduced by the proposed pass logic.





 TABLE I

 COMPLEXITY IN N-POINT FFT UNIT

	FFT Stages (n)	FFT points (N=2^n)	Total multipliers (N/2log2N)	Multipliers used in nth stage (2 ⁿ -1)
	3	8	12	4
1	4	16	32	8
	5	32	80	16
	6	64	192	32
	7	128	448	64
	8	256	1024	128
1	9	512	2304	256
	to K	est "		

Pass-logic

II.

The proposed PL is illustrated in Fig.2 eliminates the number of multipliers available at the nth stage of DIF FFT architecture. This module accepts the input as $\{0, +1,-1\}$ which can be replaced by two bit data $\{00, 01 \text{ and } 11\}$ and a TF where a and b are the real and imaginary inputs which has (w+1) bits in signed magnitude form. The truth table for PL is shown in table II for various input combinations output is obtained.

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Fig 2 R2 butterfly with PL (used only in input/nth stage)

Fig.3 illustrates the pass logic module, all the redundant adders and bit-shifters are eliminated by using PL technique. In general 'w-bit' multiplier algorithm will require (w - 1) numbers of (2w - 1) bit adders and (w - 1) number of shift operations to meet the worst data cases. Hence the power required and hardware complexity are considerably reduced.



III. RESULTS AND COMPARISONS

The experimental results of the proposed FFT software architecture are presented in this chapter. Simulation

and implementation of FFT with respect to Xilinx ISE 14.2i.are shown.











FIGURE 6: PASS-LOGIC WAVEFORM

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N-point FFT	No of LUTs utilized	Power(mw)	Delay(ns)	
256	1205	1.213	4.966	
512	1421	1.209	5.287	
1024	1684	1.214	4.199	

- Design Overview	Slice Logic Utilization	Used	Available	Utilization	Note(s)
IOB Properties	Number of Slice LUTs	1,108	69,120	1%	
Borne Contraints Traing Contraints Pinout Report Struing Contraints Struing Contraints Struing Contraints Struing Thraing Contraints Struing Struing Parser Messages Map Messages Map Messages Traints Messages	Number used as logic	1,100	69,120	1%	
	Number using O6 output only	975			
	Number using O5 output only	23			
	Number using O5 and O6	102			
	Number used as exclusive route-thru	8			
	Number of route-thrus	35			
	Number using O6 output only	27			
	Number using O5 output only	4			
	Number using OS and O6	4			
Bitgen Messages All Implementation Messages	Number of occupied Slices	303	17,280	1%	
etailed Reports	Number of LUT Flip Flop pairs used	1,108			
Synthesis Report	Number with an unused Flip Flop	1,108	1,108	100%	
Properties Enable Message Filtering	Number with an unused LUT	0	1,108	0%	
	Number of fully used LUT-FF pairs	0	1,108	0%	
Show Clock Report	Number of slice register sites lost to control set restrictions	0	69,120	0%	
Show Warnings	Number of bonded IOBs	515	640	80%	
Show Errors	IOB Flip Flops	256			
	Number of BUFG/BUFGCTRLs	1	32	3%	
	Number road as RIFGs	1			

			Total
N-point FFT	Ту <mark>ре</mark>	No of	power
_	(A)-FFT	LUTs	(mw)
	(B)-FFT with PL	utilized	
4	(A)	5	1.186
	(B)	4	1.182
8	(A)	8	1.201
	(B)	6	1.198
16	(A)	13	1.221
	(B)	11	1.204

00

01 11

11

01

FIGURE 8: SYNTHESIS REPORT OF 8 POINT DIF WITH PASS LOGIC

The	implementation	results	done	on X	Cilinx	Virtex-5,
хсэ	LX110T and th	e c <mark>ompar</mark>	ison pa	ramete	rs suc	h as area,
powe	r and delay has	to be a	nalyzed	. Table	e III s	shows the
desig	summary of	N-point I	FFT. Ta	able IV	V com	pares the
desig	n parameter with	normal F	FT and	FFT w	ith pas	s logic.

TABLE III

DESIGN SUMMARY- COMPARISON OF N-POINT FFT

TABLE IV

COMPARISON OF N-POINT FFT WITH PASS LOGIC

FIGURE 7: 8 POINT FFT WITH PASS LOGIC

X1: 1,000,000 ps

dk

in10.0

in3/1-0

in4(1-0

IV. CONCLUSION AND FUTURE ENHANCEMENT

Design of FFT and IFFT was proposed and the FFT algorithm was successfully simulated and tested through Xilinx 14.2i. An 8-point FFT and IFFT with pass logic were computed. This modification reduces the number of multipliers requires in the FFT architecture, in order to reduce area, power and hardware complexity. This is an encouraging phenomenon for high performance OFDM-based wireless communication systems.

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